

# BOSEM Noise Analysis

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Ⓟ ⇒ must verify      ✓ ⇒ checked

## Analysis of BOSEM Performance per "Sensors and Actuators for the Advanced LIGO Mirror Suspensions" P1100208

Ⓟ Per page 4, the BOSEM yields  $3 \times 10^{-10} \frac{m}{\sqrt{Hz}}$  sensitivity from 0.1 → 10 Hz

✓ TYP0 Per Figure 6, the BOSEM sensitivity should be:

$9 \times 10^{-10} \frac{m}{\sqrt{Hz}}$  @ 0.1 Hz  
 $3 \times 10^{-10} \frac{m}{\sqrt{Hz}}$  @ 1 Hz  
 $1 \times 10^{-10} \frac{m}{\sqrt{Hz}}$  @ 10 Hz

Ⓟ Per page 6, the mean photocurrent at  $\frac{1}{2}$  light is  $62.5 \mu A$   
↳ The PD responsivity is  $0.55 \frac{A}{W}$

$V_{out}$  at  $\frac{1}{2}$  light =  $\sim 8.5 VDC$  per Figure 6

Transimpedance was ⇒  $160 K\Omega \times$  factor of 2 for diff driver =  $320 K\Omega$

Photocurrent ( $I_{PD}$ ) =  $\frac{8.5 VDC}{320 K\Omega} = 26.6 \mu A$  NOT  $62.5 \mu A$

However the FULL light photocurrent is =  $62.5 \mu A = \frac{20V}{320 \Omega}$

Ⓟ Per page 6, measuring range =  $\sim 700 \mu m$ , and the average slope of Voltage output VS position is  $\sim 20 \frac{kV}{m}$  (actual  $\frac{16V}{700 \times 10^{-6} m} = 22.9 \frac{kV}{m}$ )

Ⓟ Per page 6, the BOSEM noise performance should be limited by shot noise at  $7 \times 10^{-11} \frac{m}{\sqrt{Hz}}$  for Freq > 10 Hz. Freq < 10 Hz, the performance should be limited by  $\frac{1}{f}$  photo current noise in the LED

Shot noise at  $62.5 \mu A = \sqrt{2e I_{PD}} = 4.47 \text{ pA} / \sqrt{Hz}$

Shot noise at  $26.6 \mu A = \sqrt{2e I_{PD}} = 2.92 \text{ pA} / \sqrt{Hz}$

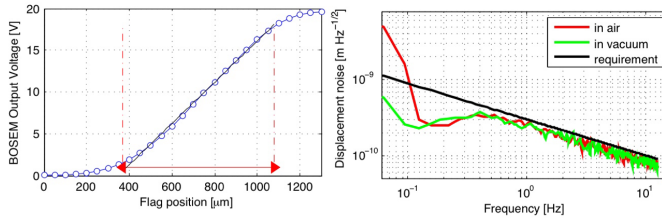
Average slope in terms of current =  $\frac{22.9 \times 10^3 V}{m} \times \frac{1}{320 K \Omega} = 7.16 \times 10^{-2} \frac{A}{m}$

So:  $4.47 \frac{pA}{\sqrt{Hz}} \times \frac{m}{7.16 \times 10^{-2} A} = 6.24 \times 10^{-11} \frac{m}{\sqrt{Hz}}$

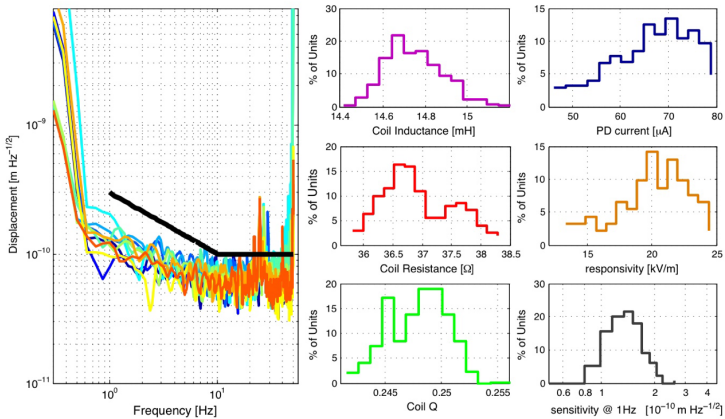
$2.92 \frac{pA}{\sqrt{Hz}} \times \frac{m}{7.16 \times 10^{-2} A} = 4.07 \times 10^{-11} \frac{m}{\sqrt{Hz}}$

Conclusion is that even at full light, shot noise equivalent displacement is  $< 1 \times 10^{-10} \frac{m}{\sqrt{Hz}}$  @ 10 Hz which is the requirement. This does not yet allow for the

other noise terms in the overall sensor (LED current noise, PD electronics noise, ADC input noise)

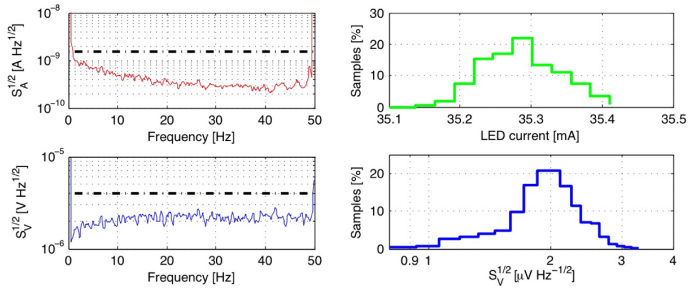


**Figure 6.** Typical performances of the BOSEMs displacement sensors. Left: response of BOSEM readout as function of position of the flag. Nominal measuring range (red lines), and fit in the linear region (black line) are also shown. Right: displacement sensitivity for a BOSEM unit, measured in-air and in-vacuum. The Advanced LIGO requirements, nominally  $3 \times 10^{-10} \text{ m Hz}^{-1/2}$  at 1 Hz and  $1 \times 10^{-10} \text{ m Hz}^{-1/2}$  at 10 Hz [24], are shown for comparison.



**Figure 7.** Typical performances of the BOSEM production articles. Left panel: displacement noise spectra of ten BOSEM samples, compared with the Advanced LIGO requirement (black line). The spikes between 20 Hz and 50 Hz are mains-related pick-up and are artefacts of the measurement system. Central column: some statistics on the reproducibility of the BOSEMs electrical properties for  $\sim 600$  BOSEM articles: inductance (nominal design value 14.7 mH), resistance ( $37.6 \Omega$ ) and electrical  $Q$  ( $243 \times 10^{-3}$ ), to be compared with  $\pm 5\%$  tolerances from the requirements [17]. Right column: distribution of the PD currents measured from  $\sim 600$  units (nominal design value  $62.5 \mu\text{A}$ , tolerance  $\pm 28\%$ ), and some statistics on responsivity and displacement noise at 1 Hz measured for the fully characterised BOSEM units (about  $\sim 20\%$  of the total).

# Satellite Box Details (D0901284)



**Figure 8.** Satellite boxes typical performances. Left, LED current supply (top) and PD readout voltage noise performances (bottom) compared to requirements (black dashed lines). Right, distribution of the LED current (top, tolerance  $\pm 5\%$ ) and PD amplifier voltage noise at 10 Hz (bottom) measured over  $\sim 230$  Satellite box units ( $\sim 920$  channels).

Per page 9, the LED current source produces 35 mA for each LED with  $\frac{0.5 \text{ nA}}{\sqrt{1\text{Hz}}}$  at 10 Hz corresponding to  $3 \times 10^{-11} \frac{\text{m}}{\sqrt{1\text{Hz}}}$  which should be  $\sim 3X <$  the requirement

① Verified  $I_{\text{LED}} = 35 \text{ mA}$

② Verified  $I_{\text{LED}}$  noise @ 10 Hz per figure 8 is  $\leq \frac{0.5 \text{ nA}}{\sqrt{1\text{Hz}}}$  @ 10 Hz

assume  $I_{\text{LED}}$  of 35 mA produces  $62.5 \mu\text{A} \Rightarrow \text{Gain} = \frac{62.5 \times 10^{-6}}{35 \times 10^{-3}} = 1.79 \times 10^{-3}$

so  $\frac{0.5 \text{ nA}}{\sqrt{1\text{Hz}}}$  becomes  $\frac{0.5 \text{ nA}}{\sqrt{1\text{Hz}}} \times 1.79 \times 10^{-3} = 8.95 \times 10^{-13} \frac{\text{A}}{\sqrt{1\text{Hz}}}$

which equates to a displacement noise of  $8.95 \times 10^{-13} \frac{\text{A}}{\sqrt{1\text{Hz}}} \times \frac{\text{m}}{7.16 \times 10^{-2} \text{ A}}$   
 $= 1.25 \times 10^{-11} \frac{\text{m}}{\sqrt{1\text{Hz}}}$

\* measured aLIGO unit  
 current noise to be  $0.7 \text{ nA}/\sqrt{1\text{Hz}}$  @ 10 Hz  
 S1000276 Unit from LHO  
 BOSEM Pedigree unknown

Gain of sensor in  
 A/m based  
 on slope =  
 $\frac{22.86 \text{ kV}}{\text{m}}$   
 $\neq 320 \text{ k}\Omega$

Per page 9, the PD amplifier [Transimpedance amplifier signal chain] converts the PD current into a  $\pm 10\text{V}$  signal with intrinsic voltage noise [interpreted to mean electronics chain dark noise]

$$< \frac{4\mu\text{V}}{\sqrt{\text{Hz}}} \text{ which equates to a displacement noise of } 5 \times 10^{-11} \frac{\text{m}}{\sqrt{\text{Hz}}} \text{ at } 10\text{Hz}$$

• Using the sensor slope of  $7.16 \times 10^{-2} \frac{\text{A}}{\text{m}} \Rightarrow$

TIA opamp current noise -  $0.2 \text{ pA}/\sqrt{\text{Hz}}$ ,  $1/f$  corner  $\sim 400\text{Hz}$

voltage noise -  $7.9 \text{ nV}/\sqrt{\text{Hz}}$  (negligible)

$160\text{k}\Omega$  feedback resistor  $I_{\text{noise}} = 0.32 \text{ pA}/\sqrt{\text{Hz}}$

Input referred TIA electronics noise =  $\sqrt{(0.2)^2 + (0.32)^2} = \sim 0.4 \text{ pA}/\sqrt{\text{Hz}}$

$$\text{Equivalent displacement noise} = \frac{0.4 \text{ pA}}{\sqrt{\text{Hz}}} \cdot \frac{\text{m}}{7.16 \times 10^{-2} \text{ A}} = \boxed{5.59 \times 10^{-12} \frac{\text{m}}{\sqrt{\text{Hz}}}}$$

• If the shot noise at  $62.5 \mu\text{A } I_{\text{pd}}$  is included in the electronics noise, the shot noise ( $4.47 \text{ pA}/\sqrt{\text{Hz}}$ ) will dominate over the TIA noise ( $0.4 \text{ pA}/\sqrt{\text{Hz}}$ ) and yield an equivalent displacement noise of:

$$4.47 \times 10^{-12} \frac{\text{A}}{\sqrt{\text{Hz}}} \cdot \frac{\text{m}}{7.16 \times 10^{-2} \text{ A}} = \boxed{6.24 \times 10^{-11} \frac{\text{m}}{\sqrt{\text{Hz}}}}$$

• If  $I_{\text{pd}}$  is taken at  $1/2$  light, then the equivalent displacement noise is:

$$2.92 \times 10^{-12} \frac{\text{A}}{\sqrt{\text{Hz}}} \cdot \frac{\text{m}}{7.16 \times 10^{-2} \text{ A}} = \boxed{4.08 \times 10^{-11} \frac{\text{m}}{\sqrt{\text{Hz}}}}$$

• The quoted  $4\mu\text{V}/\sqrt{\text{Hz}}$  is translated to  $\text{m}/\sqrt{\text{Hz}}$  at  $10\text{Hz}$  by knowing the whitening gain @  $10\text{Hz}$  (19) & the trans Z ( $320\text{k}\Omega$ )  $\therefore$

$$4 \times 10^{-6} \frac{\text{V}}{\sqrt{\text{Hz}}} \times \frac{\text{A}}{320 \times 10^3 \cdot 19} \cdot \frac{\text{m}}{7.16 \times 10^{-2} \text{ A}} = \boxed{9.19 \times 10^{-12} \frac{\text{m}}{\sqrt{\text{Hz}}}}$$

Conclusion: We do not understand how  $4\mu\text{V}/\sqrt{\text{Hz}}$  equates to  $5 \times 10^{-11} \text{ m}/\sqrt{\text{Hz}}$

The whitening gain consists of

$$\text{Zero} = 0.4 \text{ Hz}$$

$$\text{Pole} = 10 \text{ Hz}$$

$$\text{DC Gain} = -1$$

$$\left( \begin{array}{l} \text{ACTUAL} \\ \text{ZERO} = 0.3835 \text{ Hz} \\ \text{POLE} = 10.61 \text{ Hz} \end{array} \right)$$

∴ if one were to use the pole & zero to calculate the gain at 10 Hz, one would conclude:

$$\text{Gain @ 10 Hz} = \frac{10}{0.4} = 25 \quad \text{Now using this} \Rightarrow$$

Output noise quoted as  $4 \mu\text{V}/\sqrt{\text{Hz}}$ , DC Gain of sensor =  $\sqrt{20 \times 10^3} \frac{\text{V}}{\text{m}}$  quoted in note

$$\text{So: } \frac{4 \mu\text{V}}{\sqrt{\text{Hz}}} \cdot \frac{\text{m}}{20 \times 10^3 \text{ V}} \Rightarrow 2 \times 10^{-10} \frac{\text{m}}{\sqrt{\text{Hz}}}$$

Translating the 1 Hz & 10 Hz sensing noise specs to equivalent motion for inclusion in a noise budget presented in units of voltage noise:

$$\text{at 1 Hz spec is } 3 \times 10^{-10} \text{ m}/\sqrt{\text{Hz}}$$

$$10 \text{ Hz} \quad " \quad " \quad 1 \times 10^{-10} \text{ m}/\sqrt{\text{Hz}}$$

The DC gain from motion to output voltage is quoted in P

$$\text{as } 700 \mu\text{m} \Rightarrow 16 \text{ VDC} \quad \text{so} \quad \frac{16 \text{ VDC}}{700 \text{e-6 m}} = 22.86 \frac{\text{KV}}{\text{m}} \quad \left( \begin{array}{l} \text{using } 160 \text{ k}\Omega \text{ TIA} \\ \text{resistor \& diff dr.} \\ \text{gain} = 2 \end{array} \right)$$

A shift from 160 kΩ to 120 kΩ for TIA  $\Rightarrow$

$$22.86 \frac{\text{KV}}{\text{m}} \times \frac{120 \text{ k}\Omega}{160 \text{ k}\Omega} = \boxed{17.14 \frac{\text{KV}}{\text{m}}}$$

So to predict output voltage noise at 1 Hz & 10 Hz this slope in conjunction with the whitening gain (2.87 @ 1 Hz, 19 @ 10 Hz) is given by

$$\frac{1 \text{ Hz}}{120 \text{ k}\Omega} \quad 3 \times 10^{-10} \frac{\text{m}}{\sqrt{\text{Hz}}} \times 17.14 \frac{\text{KV}}{\text{m}} \times 2.87 = 1.48 \times 10^{-5} \frac{\text{V}}{\sqrt{\text{Hz}}}$$

$$\frac{10 \text{ Hz}}{120 \text{ k}\Omega} \quad 1 \times 10^{-10} \times 17.14 \frac{\text{KV}}{\text{m}} \times 19 = 3.26 \times 10^{-5} \frac{\text{V}}{\sqrt{\text{Hz}}}$$

